<span id="page-0-0"></span>Lecture 13 – ME6402, Spring 2025 Time-Varying Systems and Lyapunov Design

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#### Goals of Lecture 13

- ▶ Linear Time-Varying Systems
- Differential Lyapunov Equation
- ▶ Lyapunov Design Examples
- Additional Reading
	- Khalil Chapter 4.6

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$$
\dot{x} = A(t)x \qquad x(t) = \Phi(t, t_0)x(t_0)
$$

 $\blacktriangleright$  The state transition matrix  $\Phi(t,t_0)$  satisfies the equations:

$$
\frac{\partial}{\partial t}\Phi(t,t_0) = A(t)\Phi(t,t_0)
$$
  

$$
\frac{\partial}{\partial t_0}\Phi(t,t_0) = -\Phi(t,t_0)A(t_0)
$$

<span id="page-1-0"></span>▶ Khalil Section 4.6, Sastry Section 5.7

► No eigenvalue test for stability in the time-varying case:  
\n
$$
A(t) = \begin{bmatrix} -1 + 1.5 \cos^2 t & 1 - 1.5 \sin t \cos t \\ -1 - 1.5 \sin t \cos t & -1 + 1.5 \sin^2 t \end{bmatrix}
$$
\neigenvalues: -0.25≠i0.25√7 for all *t*, but unstable:  
\n
$$
\Phi(t,0) = \begin{bmatrix} e^{0.5t} \cos t & e^{-t} \sin t \\ e^{-0.5t} \sin t & e^{-t} \cos t \end{bmatrix}
$$

▶ Khalil Section 4.6, Sastry Section 5.7

▶ For linear systems uniform asymptotic stability is equivalent to uniform exponential stability: Theorem:  $x = 0$  is uniformly asymptotically stable if and only if

$$
\|\Phi(t,t_0)\| \leq k e^{-\lambda(t-t_0)} \quad \text{for some} \quad k > 0, \ \lambda > 0.
$$

▶ Last lecture:  $V(t, x) = x^T P(t)x$  proves uniform exp. stability if

(i) 
$$
\dot{P}(t) + A^T(t)P(t) + P(t)A(t) = -Q(t)
$$
  
\n(ii)  $0 < k_1 I \le P(t) \le k_2 I$   
\n(iii)  $0 < k_3 I \le Q(t)$  for all *t*.

▶ Khalil Thm. 4.11, Sastry Thm. 5.33

The converse is also true:

Theorem: Suppose  $x = 0$  is uniformly exponentially stable,  $A(t)$ is continuous and bounded,  $Q(t)$  is continuous and symmetric, and there exist  $k_3, k_4 > 0$  such that

 $0 < k<sub>3</sub>I < O(t) < k<sub>A</sub>I$  for all *t*.

Then, there exists a symmetric  $P(t)$  satisfying (i)–(ii):

 $P(t) + A^T(t)P(t) + P(t)A(t) = -Q(t)$ 

(ii)  $0 < k_1 I < P(t) < k_2 I$ 

- $\blacktriangleright$  For stable linear systems, there always exists quadratic Lyapunov functions.
- $\blacktriangleright$  Find them by choosing any positive definite  $Q(t)$  and solve (differential) Lyapunov equation.

[Lecture 13 Notes – ME6402, Spring 2025](#page-0-0) 5/15

Proof:  $\overline{\mathsf{Time}}$ -invariant:  $P = \int_{0}^{\infty}$  $\boldsymbol{0}$  $e^{A^T \tau} Qe^{A \tau} d\tau$ Time-varying:  $P(t) = \int_{0}^{\infty} \Phi^{T}(\tau, t) Q(\tau) \Phi(\tau, t) d\tau$ Using the Leibniz rule, property [\(2\)](#page-1-0), and  $\Phi(t,t) = I$  we obtain:  $\dot{P}(t) = \int_t^{\infty}$  ∂  $\frac{\partial}{\partial t}\Phi^T(\tau,t)Q(\tau)\Phi(\tau,t)+\Phi^T(\tau,t)Q(\tau)\frac{\partial}{\partial t}$  $\frac{\partial}{\partial t}$ Φ(τ, t)  $\bigg)$  dτ  $-\Phi^T(t,t)Q(t)\Phi(t,t)$  $=$   $\int^{\infty}$ *t*  $\left(-A^T(t)\Phi^T(\tau,t)Q(\tau)\Phi(\tau,t)-\Phi^T(\tau,t)Q(\tau)\Phi(\tau,t)A(t)\right)dt$  $-\Phi^T(t,t)Q(t)\Phi(t,t)$  $= -A^T(t)P(t) - P(t)A(t) - Q(t).$ 

- ▶ Adaptive Control (this lecture)
- ▶ Backstepping (next lecture)
- ▶ Control Lyapunov Functions (later)

#### Adaptive Control with an Unknown Parameter

Consider

$$
\dot{y} = a^*y + u.
$$

 $\frac{Goal}{.}$  Stabilize the origin even when  $a^*$  is unknown.

- ▶ Can this be achieved with linear feedback, *u* = −*Ky* for some *K*?
- $\triangleright$  Can this be achieved with static nonlinear feedback,  $u = k(y)$ ?

#### Adaptive Control with an Unknown Parameter

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▶ Can this be achieved with linear feedback, *u* = −*Ky* for some *K*?

A: Not unless an a priori bound on  $|a^*|$  is known.

 $\triangleright$  Can this be achieved with static nonlinear feedback.  $u = k(y)$ ? A: Try  $u = -ky^3$ ,  $k > 0$ . If  $a^* > 0$ , this introduces two new, stable equilibria at  $\pm \sqrt{a^*/k}$ . Trajectories, therefore remain bounded, but still no stability at origin.

#### Adaptive Control with an Unknown Parameter, Dynamic Feedback

Let's try building a dynamic feedback controller.

- $\triangleright$  Dynamic means the controller itself has a state variable, and therefore memory.
- $\blacktriangleright$  Let the controller estimate *a* with  $\hat{a}$ .
- Goal: Design update law for *a* and controller  $u(y, \hat{a})$  to stabilize origin.
	- Our approach is Lyapunov-based: we choose a Lyapunov function candidate and work to make it an actual Lyapunov function

▶ System:

 $\dot{y} = a^*y + u,$ 

*a* ∗ unknown.

$$
\dot{y} = a^*y + u(y, \hat{a}), \qquad \dot{\hat{a}} = ??
$$

State variables are *y*,  $\hat{a}$ . Try

$$
V(y, \hat{a}) = \frac{1}{2}y^2 + \frac{1}{2}(\hat{a} - a^*)^2
$$

$$
\dot{y} = a^*y + u(y, \hat{a}), \qquad \dot{a} = ??
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State variables are *y*,  $\hat{a}$ . Try

$$
V(y, \hat{a}) = \frac{1}{2}y^2 + \frac{1}{2}(\hat{a} - a^*)^2
$$
  
\n
$$
\implies \dot{V} = y\dot{y} + (\hat{a} - a^*)\dot{a} = y(a^*y + u) + (\hat{a} - a^*)\dot{a}
$$
  
\n
$$
= a^*(y^2 - \dot{a}) + uy + \hat{a}\dot{a}
$$

▶ Want *V*˙ ≤ −*y* 2 (for example) so that *y* → 0 by LaSalle

▶ *u* and  $\dot{a}$  can be functions of  $\hat{a}$  and *y*, but not  $a^*$ 

• Therefore, we choose 
$$
\hat{a} = y^2
$$
 and then  $\dot{V} = uy + \hat{a}y^2$ .  
Choose

$$
u(y, \hat{a}) = -\hat{a}y - y
$$

[Lecture 13 Notes – ME6402, Spring 2025](#page-0-0)  $10/15$ 

## Adaptive Control with an Unknown Parameter, Dynamic Feedback (cont.)

Final controlled system:

$$
\dot{y} = a^* y + u(y, \hat{a}) = (a^* - \hat{a} - 1)y
$$

$$
\dot{a} = y^2
$$

Lyapunov function  $V(y,\hat{a}) = \frac{1}{2}y^2 + \frac{1}{2}$  $\frac{1}{2}(\hat{a}-a^*)^2$  gives  $\dot{V}=-y^2$ . Apply LaSalle.

▶ Does  $y \rightarrow 0$ ?

▶ Does  $\hat{a} \rightarrow a^*$ ?

## Adaptive Control with an Unknown Parameter, Dynamic Feedback (cont.)

Final controlled system:

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- ▶ Does  $y \rightarrow 0$ ? A: Yes
- ▶ Does  $\hat{a} \rightarrow a^*$ ?

A: Not necessarily. For example,  $(0, \hat{a})$  is an equilibrium for any  $\hat{a}$ . We achieve asymptotic stability of system, but not necessarily estimator convergence

#### Model Reference Adaptive Control

Illustrated on same first order system:

 $\dot{y} = a^*y + u$ 

where  $a^*$  is unknown.

Reference model:

<span id="page-14-0"></span> $\dot{y}_m = -ay_m + r(t)$  *a* > 0, *r(t)* : reference signal.

Goal: Design a controller that guarantees  $y(t) - y_m(t) \rightarrow 0$  without the knowledge of  $a^*$ .

#### Model Reference Adaptive Control (cont.)

If we knew *a* ∗ , we would choose:

$$
u = -(a^* + a)y + r(t) \quad \Rightarrow \quad y = -ay + r(t).
$$

The tracking error  $e(t) := y(t) - y_m(t)$  then satisfies:

$$
\dot{e} = -ae \Rightarrow e(t) \rightarrow 0
$$
 exponentially.

Adaptive design when  $a^*$  (therefore,  $k^*$ ) is unknown:

$$
u = -k(t)y + r(t)
$$

where  $k(t)$  is to be designed. Then:

$$
\dot{e} = \dot{y} - \dot{y}_m = a^*y - k(t)y + ay_m = -ae - \underbrace{(k(t) - k^*)}_{=: \tilde{k}(t)}y
$$

where adding and subtracting *ay* gives the final equality.

#### ▶

$$
\dot{y} = a^*y + u
$$
  

$$
\dot{y}_m = -ay_m + r(t) \ a > 0,
$$

 $r(t)$ : reference signal.

#### Model Reference Adaptive Control (cont.)

Use the Lyapunov function:  $V = \frac{1}{2}$  $\frac{1}{2}e^2 + \frac{1}{2}$  $\frac{1}{2}\tilde{k}^2$ :  $\dot{V} = -ae^2 - \tilde{k}ey + \tilde{k}\tilde{k}$  $= -ae^2 + \tilde{k}(\dot{\tilde{k}} - ey).$ Note  $\dot{\tilde{k}} = \dot{k}$  and choose  $\dot{\tilde{k}} = ey$  so that  $\dot{V} = -ae^2$ . This guarantees stability of  $(e,\tilde{k})=(0,0)$  and boundedness of  $(e(t), \tilde{k}(t))$  since the level sets of  $V = \frac{1}{2}$  $\frac{1}{2}e^2 + \frac{1}{2}$  $\frac{1}{2}\tilde{k}^2$  are positively invariant. In addition, if  $r(t)$  is bounded, then  $y_m(t)$  in [\(15\)](#page-14-0) is bounded, and so is  $y(t) = y_m(t) + e(t)$ .

▶

$$
\dot{y}_m = -ay_m + r(t) \ a > 0,
$$

*r*(*t*) : reference signal.

#### Model Reference Adaptive Control (cont.)

Then we can apply the Theorem from Lecture 12, page 5, to the time-varying model

$$
\dot{e} = -ae - y(t)\tilde{k}, \quad \dot{\tilde{k}} = y(t)e,
$$

and conclude from  $\dot{V}=-ae^2$  that  $e(t)\rightarrow 0.$ 

Whether  $\tilde{k}(t) \rightarrow 0 \,\, (k(t) \rightarrow k^*)$  depends on further properties of the reference signal  $r(\cdot)$  that are beyond the scope of this lecture.

▶

 $\dot{y}_m = -a y_m + r(t) \ a > 0,$ 

 $r(t)$  : reference signal.