ME 6402 – Lecture 19 feedback linearization 4 (feedback linearization for mimo systems)

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Overview:

- Frobenius Theorem continued
- Feedback linearization for MIMO Systems

Additional Reading:

• Khalil, Chapter 13.3

Full State Feedback Linearization

Recall that our condition for r = n is:

$$\begin{split} L_g h(x) &= L_{\mathrm{ad}_f \, g} h(x) = \dots = L_{\mathrm{ad}_f^{n-2} g} h(x) = 0 \text{ in a nbhd of } x_0 \quad \mbox{ (1)} \\ L_{\mathrm{ad}_f^{n-1} g} h(x_0) \neq 0. \quad \mbox{ (2)} \end{split}$$

Note: (1) can be rewritten as:

$$L_g h(x) = L_{[f,g]} h(x) = \dots = L_{[f,[f,\dots,[f,g]]]} h(x) = 0$$

where the benefit is that the h(x) term can be moved outside:

$$\frac{\partial h}{\partial x} \begin{bmatrix} g & \operatorname{ad}_f g & \operatorname{ad}_f^2 g & \dots & \operatorname{ad}_f^{n-2} g \end{bmatrix}$$

Theorem: Full-state Feedback Linearizable. The system $\dot{x} = f(x) + g(x)u$ is feedback linearizable around x_0 if and only if the following two conditions hold:

- C1) $[g(x_0) \text{ ad}_f g(x_0) \dots \text{ ad}_f^{n-1}g(x_0)]$ has rank n
- C2) $\Delta(x) = \operatorname{span}\{g(x), \operatorname{ad}_f g(x), \dots, \operatorname{ad}_f^{n-2} g(x)\}$ is involutive in a neighborhood of x_0 .

Proof. (if) Given C1 and C2 show that there exists h(x) satisfying (1)-(2).

 $\Delta(x)$ is nonsingular by C1 and involutive by C2. Thus, by the Frobenius Theorem ¹, there exists h(x) satisfying (1) and $dh(x) \neq 0$.

To prove (2) suppose, to the contrary, $L_{\mathrm{ad}_{f}^{n-1}}h(x_{0})=0$. This implies

$$dh(x_0)\underbrace{[g(x_0) \quad \text{ad}_f g(x_0) \quad \dots \quad \text{ad}_f^{n-1}g(x_0)]}_{\text{nonsingular by } C_1} = 0.$$

¹ Recall that the Frobenius Theorem states that a nonsingular distribution is completely integrable if and only if it is involutive. And recall that completely integrable tells us that there must exist n - k functions such that $\frac{\partial \phi_i}{\partial x} f_j = 0$ and $\frac{\partial \phi_i}{\partial x}$ are linearly independent

 $\begin{array}{rcl} L_{\mathrm{ad}_f\,g}h(x) &=& L_{[f,g]}h(x) \\ &=& \frac{\partial h}{\partial x} \left(\frac{\partial g}{\partial x}f(x) - \frac{\partial f}{\partial x}g(x) \right) \end{array}$

Thus $dh(x_0) = 0$, a contradiction.

(only if) Given that y = h(x) with r = n exists, that is (1)-(2) hold, show that C1 and C2 are true.

We will use the following fact² which holds when r = n:

$$L_{\mathrm{ad}_{f}^{i}g}L_{f}^{j}h(x) = \begin{cases} 0 & \text{if } i+j \le n-2\\ (-1)^{n-1-j}L_{g}L_{f}^{n-1}h(x) \ne 0 & \text{if } i+j=n-1. \end{cases}$$
(3)

Define the matrix

$$M = \begin{bmatrix} dh \\ dL_fh \\ \vdots \\ dL_f^{n-1}h \end{bmatrix} \begin{bmatrix} g & -\operatorname{ad}_f g & \operatorname{ad}_f^2 g & \dots & (-1)^{n-1}\operatorname{ad}_f^{n-1}g \end{bmatrix}$$
(4)

and note that the (k, ℓ) entry is:

$$\begin{split} M_{k\ell} &= dL_f^{k-1}h(x)(-1)^{\ell-1}\operatorname{ad}_f^{\ell-1}g(x) \\ &= (-1)^{\ell-1}L_{\operatorname{ad}_f^{\ell-1}g}L_f^{k-1}h(x). \end{split}$$

Then, from (3):

$$M_{k\ell} = \begin{cases} 0 & \ell + k \le n \\ \neq 0 & \ell + k = n + 1. \end{cases}$$

Since the diagonal entries are nonzero, M has rank n and thus the factor

$$\left[\begin{array}{ccc}g & -\operatorname{ad}_f g & \operatorname{ad}_f^2 g & \dots & (-1)^{n-1}\operatorname{ad}_f^{n-1} g\end{array}\right]$$

in (4) must have rank n as well. Thus C1 follows.

This also implies $\Delta(x)$ is nonsingular; thus, by the Frobenius Thm,

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complete integrability \equiv involutivity.
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 $\Delta(x)$ is completely integrable since h(x) satisfying (1) exists by assumption; thus, we conclude involutivity (C2).

Example:

Consider the following system:

$$\dot{x}_1 = x_2 + 2x_1^2$$

 $\dot{x}_2 = x_3 + u$
 $\dot{x}_3 = x_1 - x_3$

Feedback linearizability was shown on page 1 by inspection: $y = x_3$ gives relative degree = 3. Can we verify this choice of y with the



theorem above? We will begin by computing the elements of the span³:

$$f(x) = \begin{bmatrix} x_2 + 2x_1^2 \\ x_3 \\ x_1 - x_3 \end{bmatrix} g(x) = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}$$
$$[f,g](x) = \begin{bmatrix} -1 \\ 0 \\ 0 \end{bmatrix} [f,[f,g]](x) = \begin{bmatrix} 4x_1 \\ 0 \\ 1 \end{bmatrix}$$

³ Recall that $[f,g] = \frac{\partial g}{\partial x}f(x) - \frac{\partial f}{\partial x}g(x)$ $[f,g] = 0 - \begin{bmatrix} 4x_1 & 1 & 0\\ 0 & 0 & 1\\ 1 & 0 & -1 \end{bmatrix} \begin{bmatrix} 0\\ 1\\ 0\\ 0 \end{bmatrix} = \begin{bmatrix} -1\\ 0\\ 0\\ 0 \end{bmatrix}$ $[f,[f,g]] = 0 - \begin{bmatrix} 4x_1 & 1 & 0\\ 0 & 0 & 1\\ 1 & 0 & -1 \end{bmatrix} \begin{bmatrix} -1\\ 0\\ 0\\ 0 \end{bmatrix} = \begin{bmatrix} 4x_1\\ 0\\ 1 \end{bmatrix}$

Conditions of the theorem:

1.
$$\begin{bmatrix} 0 & -1 & 4x_1 \\ 1 & 0 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$
 full rank
2.
$$\Delta = \operatorname{span} \left\{ \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}, \begin{bmatrix} -1 \\ 0 \\ 0 \end{bmatrix} \right\}$$
 involutive

$$\frac{\partial h}{\partial x} \begin{bmatrix} 0 & -1 \\ 1 & 0 \\ 0 & 0 \end{bmatrix}$$
 satisfied by $h(x) = x_3$.

Feedback Linearization Continued

Recall "strict feedback systems" discussed in Lecture 14:

$$\begin{aligned} \dot{x}_1 &= f_1(x_1) + g_1(x_1)x_2 \\ \dot{x}_2 &= f_2(x_1, x_2) + g_2(x_1, x_2)x_3 \\ \dot{x}_3 &= f_3(x_1, x_2, x_3) + g_3(x_1, x_2, x_3)x_4 \\ &\vdots \\ \dot{x}_n &= f_n(x) + g_n(x)u. \end{aligned}$$
(5)

Such systems are feedback linearizable when $g_i(x_1, ..., x_i) \neq 0$ near the origin, i = 1, 2, ..., n, because the relative degree is n with the choice of output $y = h(x) = x_1$:

$$y^{(n)} = L_f^n h(x) + \underbrace{g_1(x_1)g_2(x_1, x_2) \cdots g_n(x)}_{L_g L_f^{n-1} h(x) \neq 0} u.$$

Feedback linearizability is lost when $g_i(0) = 0$ for some *i*; however, backstepping may be applicable as the following example illustrates:

Example 1:

$$\begin{array}{rcl} \dot{x}_1 &=& x_1^2 x_2 \\ \dot{x}_2 &=& u. \end{array}$$

Treat x_2 as virtual control and let $\alpha_1(x_1) = -x_1$ which stabilizes the x_1 -subsystem, as verified with Lyapunov function $V_1(x_1) = \frac{1}{2}x_1^2$. Then $z_2 := x_2 - \alpha_1(x_1)$ satisfies $\dot{z}_2 = u - \dot{\alpha}_1$, and

$$u = \dot{\alpha}_1 - \frac{\partial V_1}{\partial x_1} x_1^2 - k_2 z_2 = -x_1^2 x_2 - x_1^3 - k_2 (x_2 + x_1)$$

achieves global asymptotic stability:

$$V = \frac{1}{2}x_1^2 + \frac{1}{2}z_2^2 \quad \Rightarrow \quad \dot{V} = -x_1^4 - k_2 z_2^2.$$

In contrast the system is not feedback linearizable, because condition (C1) in the theorem for feedback linearizability (Lecture 18, p.4) fails. To see this note that

$$f(x) = \begin{bmatrix} x_1^2 x_2 \\ 0 \end{bmatrix}, \quad g(x) = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad \operatorname{ad}_f g(x) = \begin{bmatrix} f, g \end{bmatrix}(x) = \begin{bmatrix} -x_1^2 \\ 0 \end{bmatrix},$$

thus, with n = 2 and $x_0 = 0$,

$$[g(x_0) \quad \operatorname{ad}_f g(x_0) \quad \dots \quad \operatorname{ad}_f^{n-1} g(x_0)] = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix},$$

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which is rank deficient.

Multi-Input Multi-Output Systems

Consider now a MIMO system with *m* inputs and *m* outputs:

$$\dot{x} = f(x) + \sum_{i=1}^{m} g_i(x)u_i = f(x) + \left[g_1(x) \dots g_m(x)\right] \begin{bmatrix} u_1 \\ \vdots \\ u_m \end{bmatrix}$$
(6)
$$y_i = h_i(x), \quad i = 1, \cdots, m.$$

Let r_i denote the number of times we need to differentiate y_i to hit at least one input. Then,

$$\begin{bmatrix} y_1^{(r_1)} \\ \vdots \\ y_m^{(r_m)} \end{bmatrix} = \underbrace{\begin{bmatrix} L_f^{r_1} h_1(x) \\ \vdots \\ L_f^{r_m} h_m(x) \end{bmatrix}}_{=: B(x)} + \underbrace{\begin{bmatrix} L_{g_1} L_f^{r_1 - 1} h_1(x) & \cdots & L_{g_m} L_f^{r_1 - 1} h_1(x) \\ \vdots & & \vdots \\ L_{g_1} L_f^{r_m - 1} h_m(x) & \cdots & L_{g_m} L_f^{r_m - 1} h_m(x) \end{bmatrix}}_{=: A(x)} \begin{bmatrix} u_1 \\ \vdots \\ u_m \end{bmatrix}.$$

If A(x) is nonsingular, then the feedback law

$$u = A(x)^{-1}(-B(x) + v)$$

input/output linearizes the system, creating m decoupled chains of integrators:

$$y_i^{(r_i)} = v_i, \quad i = 1, \dots, m.$$

We say that the system has *vector relative degree* $\{r_1, \dots, r_m\}$ if the matrix A(x) defined above is nonsingular.

Example 2: The kinematic model of a unicycle, depicted below, is

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} \cos x_3 \\ \sin x_3 \\ 0 \end{bmatrix} u_1 + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u_2,$$

where u_1 is the speed and u_2 is the angular velocity.



Let $y_1 = x_1$ and $y_2 = x_2$, and note that

$$\begin{bmatrix} \dot{y}_1\\ \dot{y}_2 \end{bmatrix} = \underbrace{\begin{bmatrix} \cos x_3 & 0\\ \sin x_3 & 0 \end{bmatrix}}_{=:A(x)} \begin{bmatrix} u_1\\ u_2 \end{bmatrix}.$$

Since A(x) is singular, the system does not have a well-defined vector relative degree.

Normal form for MIMO systems

The notion of zero dynamics and the normal form can be extended to MIMO systems⁴. If the system has vector relative degree $\{r_1, \dots, r_m\}$, ⁴se then $r := r_1 + \dots + r_m \leq n$ and

⁴ see, *e.g.*, Sastry, Section 9.3

$$\eta := [h_1(x) \ L_f h_1(x) \cdots L_f^{r_1 - 1} h_1(x) \ \cdots \ h_m(x) \ L_f h_m(x) \cdots L_f^{r_m - 1} h_m(x)]^T$$

defines a partial set of coordinates. As in normal form discussed in Lecture 17, one can find n - r additional functions $z_1(x), \dots, z_{n-r}(x)$ so that $x \mapsto (z, \eta)$ is a complete coordinate transformation.

Full-state feedback linearization amounts to finding m output functions h_1, \dots, h_m such that the system has vector relative degree $\{r_1, \dots, r_m\}$ with $r_1 + \dots + r_m = n$. Necessary and sufficient conditions for the existence of such functions, analogous to those in Lecture 18 for SISO systems, are available⁵.

Example 3: Consider the following model of a *planar vertical take-off* and landing (PVTOL) aircraft⁶

$$\begin{aligned} \ddot{x} &= -\sin(\theta)u_1 + \mu\cos(\theta)u_2 \\ \ddot{z} &= \cos(\theta)u_1 + \mu\sin(\theta)u_2 - 1 \\ \ddot{\theta} &= u_2, \end{aligned}$$

where μ is a constant that accounts for the coupling between the rolling moment and translational acceleration, and -1 in the second equation is the gravitational acceleration, normalized to unity by appropriately scaling the variables.



If we take x and z as the two outputs we get

$$\begin{bmatrix} \ddot{x} \\ \ddot{z} \end{bmatrix} = \begin{bmatrix} 0 \\ -1 \end{bmatrix} + \underbrace{\begin{bmatrix} -\sin\theta & \mu\cos\theta \\ \cos\theta & \mu\sin\theta \end{bmatrix}}_{A(\theta)} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

where $A(\theta)$ is invertible when $\mu \neq 0$:

$$A^{-1}(\theta) = \begin{bmatrix} -\sin\theta & \cos\theta \\ \frac{1}{\mu}\cos\theta & \frac{1}{\mu}\sin\theta \end{bmatrix}.$$

Thus the systems has vector relative degree $\{2, 2\}$ This implies that when $\mu \neq 0$, and the input/output linearizing controller is

$$\begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = \begin{bmatrix} -\sin\theta & \cos\theta \\ \frac{1}{\mu}\cos\theta & \frac{1}{\mu}\sin\theta \end{bmatrix} \left(\begin{bmatrix} 0 \\ 1 \end{bmatrix} + \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} \right).$$

The zero dynamics is obtained by substituting $u_2^* = \frac{1}{\mu} \sin \theta$, needed to maintain *z* at a constant value and \dot{z} at zero, in the dynamical equation for θ :

$$\ddot{\theta} = \frac{1}{\mu}\sin\theta.$$

The system is nonminimum phase for $\mu > 0$, since $\theta = 0$ is unstable.

⁵ see, e.g., Sastry, Proposition 9.16

⁶ Sastry, Section 10.4.2